Ordering number : ENA2108

SANYO Semiconductors
DATA SHEET

An ON Semiconductor Company

STK433-840N-E — Thick-Film Hybrid IC 4channel class-AB Audio Power IC 40W × 4ch

Overview

The STK433-840N-E is 4 channels class-AB audio frequency power amplifier hybrid IC.

Application

- Audio Power amplifiers

Features

- Pin-to-pin compatible outputs ranging from 40W to 80W.
- Output load impedance: R_L = 6Ω recommended.
- Allows the use of predesigned applications for standby and mute circuit.
- Miniature package.
- Allowable load shorted time: 0.3 second

Series model

	STK433-040N-E	STK433-060N-E	STK433-130N-E	STK433-330N-E
Output1 (10%/1kHz)	40W × 2ch	50W × 2ch	150W × 2ch	150W × 3ch
Output2 (0.4%/20Hz to 20kHz)	25W × 2ch	35W × 2ch	100W × 2ch	100W × 3ch
Max. rating V_{CC} (quiescent)	±38V	±46V	±71.5V	±71.5V
Max. rating V_{CC} (6Ω)	±36V	±40V	±63V	±63V
Recommended operating V_{CC} (6Ω)	±24V	±27V	±44V	±44V
Dimensions (excluding pin height)	47.0mm×25.6mm×9.0mm		67.0mm×25.6mm×9.0mm	64.0mm×36.6mm×9.0mm

	STK433-840N-E	STK433-870N-E	STK433-890N-E
Output1 (10%/1kHz)	40W × 4ch	60W × 4ch	80W × 4ch
Output2 (0.4%/20Hz to 20kHz)	25W × 4ch	40W × 4ch	50W × 4ch
Max. rating V_{CC} (quiescent)	±38V	±50V	±54V
Max. rating V_{CC} (6Ω)	±36V	±44V	±47V
Recommended operating V_{CC} (6Ω)	±25V	±30V	±34V
Dimensions (excluding pin height)	64.0mm×31.1mm×9.0mm		78.0mm×44.1mm×9.0mm

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Specifications

Absolute Maximum Ratings at Ta = 25°C, Tc = 25°C unless otherwise specified

Parameter	Symbol	Conditions	Ratings	Unit
Maximum power supply voltage	V_{CC} max (0)	Non signal	±38	V
	V_{CC} max (1)	Signal, $R_L \geq 6\Omega$	±36	V
	V_{CC} max (2)	Signal, $R_L = 4\Omega$	±30	V
Minimum operation supply voltage	V_{CC} min		±10	V
#13 Operating voltage *5	VST OFF max	#13pin voltage	-0.3 to +5.5	V
Thermal resistance	θj-c	Per power transistor	4.2	°C/W
Junction temperature	Tj max	Both the Tj max and Tc max conditions must be met.	150	°C
Operating substrate temperature	Tc max		125	°C
Storage temperature	Tstg		-30 to +125	°C
Allowable time for load short-circuit *4	ts	V_{CC} = ±25V, R_L = 6Ω, f = 50Hz P_O = 25W, 1ch drive	0.3	s

Operating Characteristics at Tc = 25°C, R_L = 6Ω (Non-inductive Load), Rg = 600Ω, VG = 30dB

Parameter	Symbol	Conditions *2					Ratings			Unit
		V_{CC} [V]	f [Hz]	P_O [W]	THD [%]		min	typ	max	
Output power *1	P_O 1	±25	20 to 20k		0.6		23	25		W
	P_O 2	±25	1k		10			40		
Total harmonic distortion *1	THD 1	±25	20 to 20k	5.0		VG=30dB			0.6	%
	THD 2	±25	1k					0.02		
Frequency characteristics *1	f_L, f_H	±25		1.0		+0 -3dB	20 to 50k			Hz
Input impedance	ri	±25	1k	1.0				55		kΩ
Output noise voltage *3	V_{NO}	±30				Rg=2.2kΩ			1.0	mVrms
Quiescent current	I_{CCO}	±30				No load	30	60	140	mA
Quiescent current at stand-by	I_{CST}	±30				VST=0V			1.0	mA
Neutral voltage	V_N	±30					-70	0	+70	mV
#13 Stand-by ON threshold *5	VST ON	±25				Stand-by		0	0.6	V
#13 Stand-by OFF threshold *5	VST OFF	±25				Operation	2.5	3.0	5.5	V

Note

*1. 1channel operation.

*2. All tests are measured using a constant-voltage supply unless otherwise specified

*3. The output noise voltage is peak value of an average-reading meter with a rms value scale (VTVM).
A regulated AC supply (50Hz) should be used to eliminate the effects of AC primary line flicker noise

*4. Allowable time for load short-circuit and output noise voltage are measured using the specified transformer power supply.

*5. The impression voltage of '#13 (Stand-By) pin' must not exceed the maximum rating.
Power amplifier operate by impressing voltage +2.5 to +5.5V to '#13 (Stand-By) pin'.

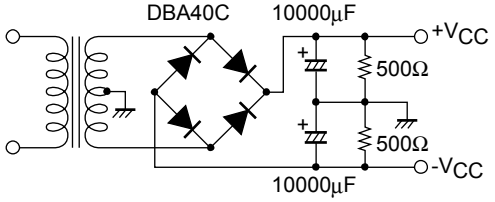
* Please connect - PreV_{CC} pin (#1 pin) with the stable minimum voltage.
and connect so that current does not flow in by reverse bias.

* In case of heat sink design, we request customer to design in the condition to have assumed market.

* The case of this Hybrid-IC is using thermosetting silicon adhesive (TSE322SX).

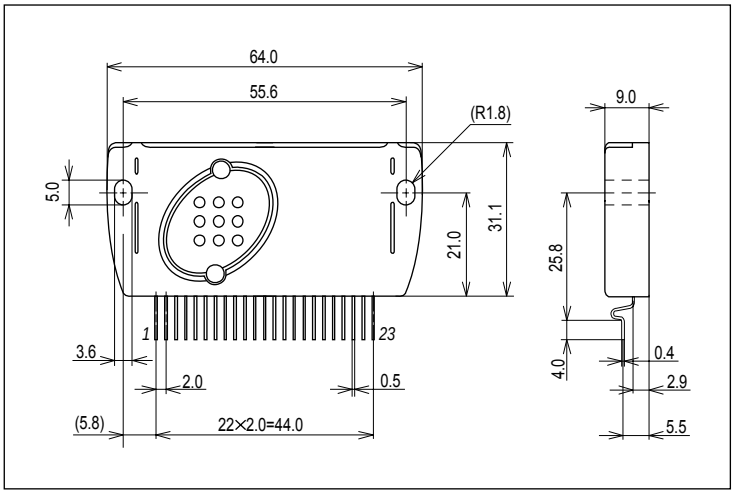
* Weight of HIC : (typ) 20.6g
Outer carton dimensions (W×L×H) : 502mm×247mm×282mm

Specified transformer power supply
(Equivalent to MG-200)



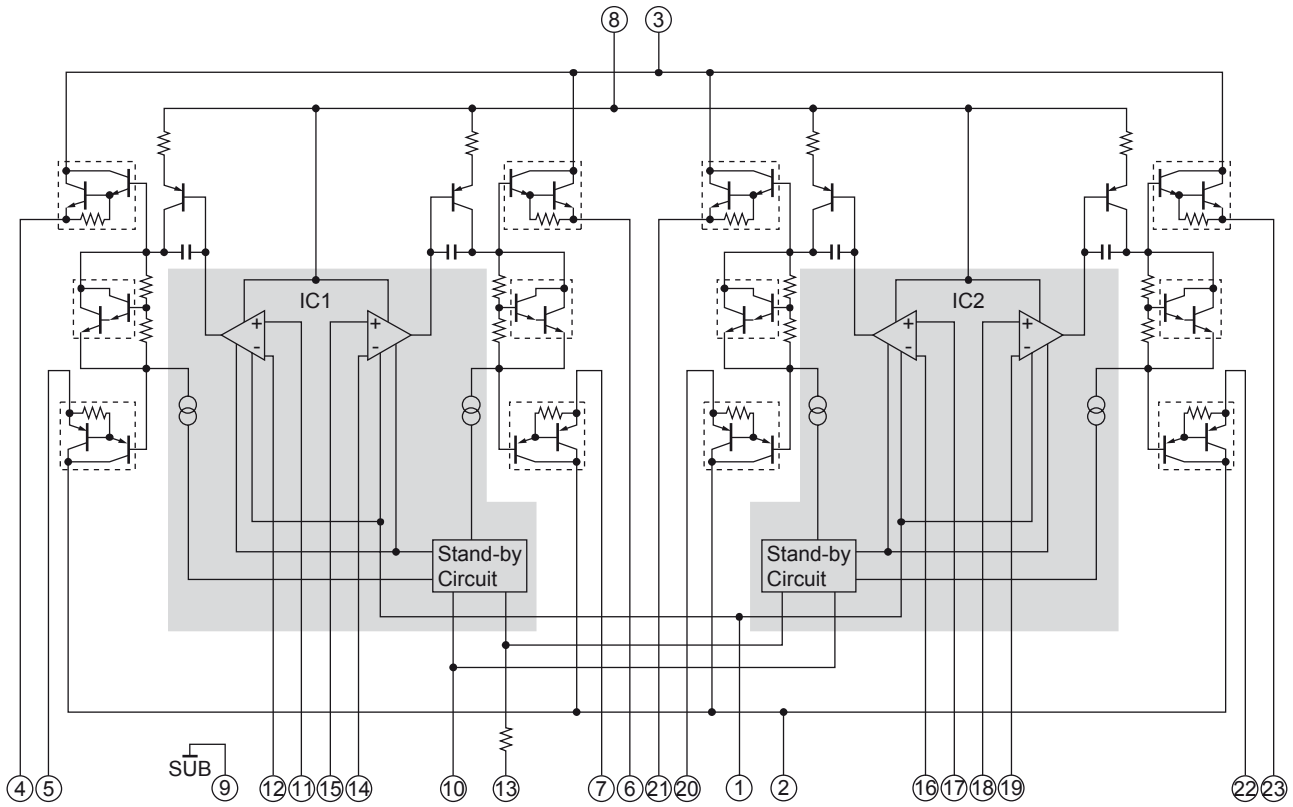
Package Dimensions

unit : mm (typ)

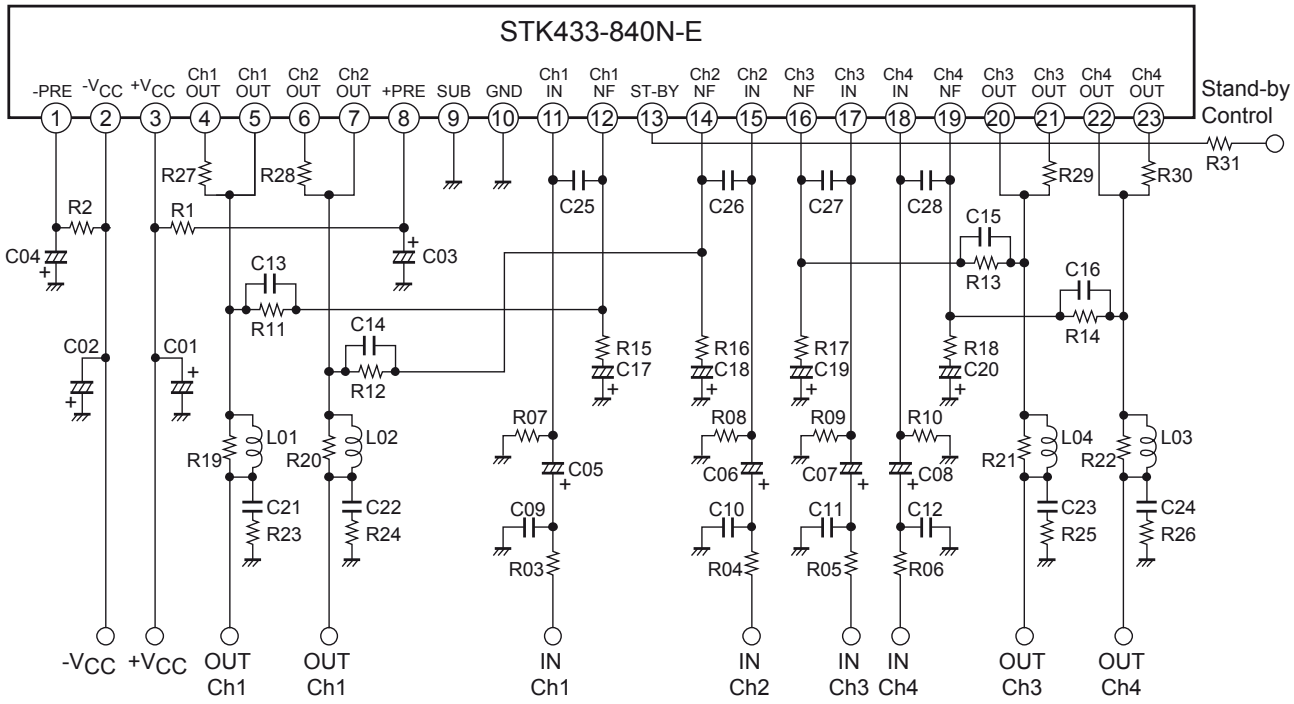


RoHS directive pass

Equivalent Circuit

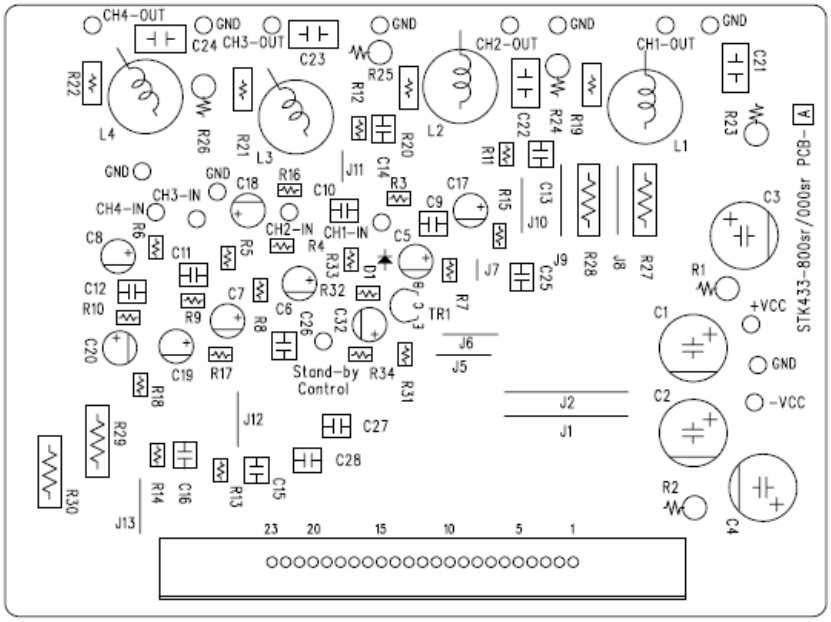


Application Circuit

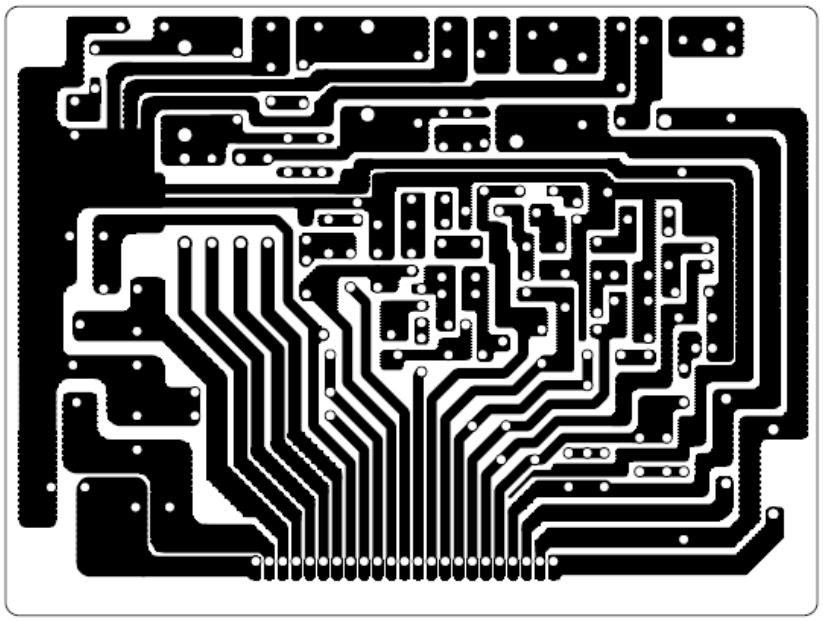


PCB Layout Example

Top view



Bottom view



STK433-800NSr PCB PARTS LIST

PCB Name : STK403-800Sr PCBA

Location No.		RATING	Component	
			STK433-840N-E/890N-E	STK433-870N-E
Hybrid IC#1 Pin Position		-	1	
R01, R02		100Ω, 1W	○	
R03, R04, R05, R06		1kΩ, 1/6W	○	
R07, R08, R09, R10, R11, R12, R13, R14		56KΩ, 1/6W	○	
R15, R16, R17, R18		1.8KΩ, 1/6W	○	
R19, R20, R21, R22		4.7Ω, 1/4W	○	
R23, R24, R25, R26		4.7Ω, 1W	○	
R27, R28, R29, R30		0.22Ω, 5W	○	
R32, R33, R34, R35		0.22Ω, 5W	-	○
C01, C02, C03, C04		100μF, 100V	○	
C05, C06, C07, C08		2.2μF, 50V	○ (*1)	
C09, C10, C11, C12		470pF, 50V	○	
C13, C14, C15, C16		***pF, 50V	3pF	
C17, C18, C19, C20		10μF, 10V	○ (*1)	
C21, C22, C23, C24		0.1μF, 50V	○	
C25, C26, C27, C28		***pF, 50V	100pF	
L01, L02, L03, L04		3μH	○	
Stand-By Control Circuit	Tr1	VCE ≥ 50V, IC ≥ 10mA	○	
	D1	Di	○	
	R31	1.8kΩ, 1/6W	○	
	R32	33kΩ, 1/6W	○	
	R33	1kΩ, 1/6W	○	
	R34	2kΩ, 1/6W	○	
	C32	33μF, 10V	○	
		-		
		-		
		-		
		-		

(*1) Capacitor mark “A” side is “ – ” (negative).

Recommended external components

STK433-840N-E/890N-E

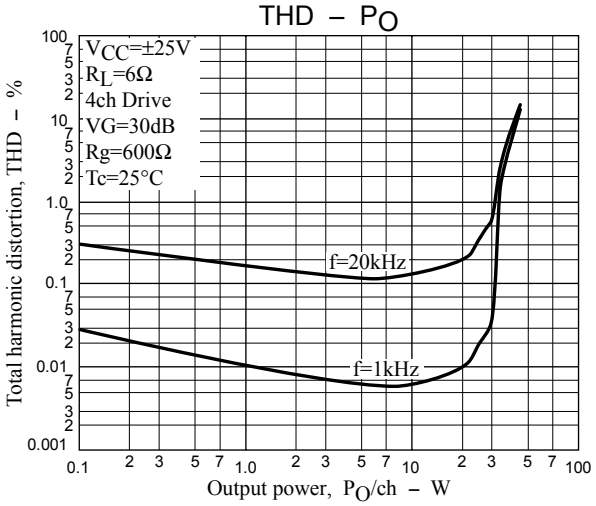
Parts Location	Recommended value	Circuit purpose	Above Recommended value	Below Recommended value
R01, R02	100Ω/1W	Resistance for Ripple filters. (Fuse resistance is recommended. Ripple filter is constituted with C03, C04.)	Short-through current may decrease at high frequency.	Short-through current may increase at high frequency.
R03,R04,R05, R06	1kΩ	Resistance for input filters.	-	-
R07,R08,R09, R10	56kΩ	Input impedance is determined.	Output neutral voltage (VN) shift. (It is referred that R07=R11, R08=R12, R09=R13, R10=R14)	
R11,R12,R13, R14	56kΩ	Voltage Gain (VG) is determined with R15, R16, R17, R18	-	-
R15,R16,R17, R18	1.8kΩ	Voltage Gain (VG) is determined with R11, R12, R13, and R14. (As for VG, it is desirable to set up by R15, R16, R17, and R18.)	It may oscillate. (Vg < 30dB)	With especially no problem
R19,R20,R21, R22	4.7Ω	Resistance for oscillation prevention.	-	-
R23,R24,R25, R26	4.7Ω/1W	Resistance for oscillation prevention.	-	-
R27,R28,R29, R30	0.22Ω ±10%, 5W	Output emitter resi@stor (Metal-plate Resistor is recommended.)	Decrease of Maximum output Power	It may cause thermal runaway
R31	Note *4	Select Restriction resistance, for the impression voltage of '#13 (Stand-By) pin' must not exceed the maximum rating.		
C01, C02	100μF/100V	Capacitor for oscillation prevention. • Locate near the HIC as much as possible. • Power supply impedance is lowered and stable operation of the IC is carried out. (Electrolytic capacitor is recommended.)	-	-
C03,C04	100μF/100V	Decoupling capacitor • The Ripple ingredient mixed in an input side Is removed from a power supply line. (Ripple filter is constituted with R01, R02.)	The change in the Ripple ingredient mixed in an input side from a power supply line	
C05,C06,C07, C08	2.2μF/50V	Input coupling capacitor. (For DC current prevention.)	-	
C09,C10,C11, C12	470pF	Input filter capacitor • A high frequency noise is reduced with the filter constituted by R03, R04, R05, R06.	-	
C13,C14,C15, C16	5pF	Capacitor for oscillation prevention.	It may oscillate.	
C17,C18,C19, C20	10μF/10V	Negative feedback capacitor. The cutoff frequency of a low cycle changes. ($fL = 1/(2\pi \cdot C17 \cdot R15)$)	The voltage gain (VG) of low frequency is extended. However, the pop noise at the time of a power supply injection also becomes large.	The voltage gain (VG) of low frequency decreases.
C21,C22,C23, C24	0.1μF	Capacitor for oscillation prevention.	It may oscillate.	
C25,C26,C27, C28	100pF	Capacitor for oscillation prevention.	It may oscillate.	
L01,L02,L03, L04	3μH	Coil for oscillation prevention.	With especially no problem	It may oscillate.

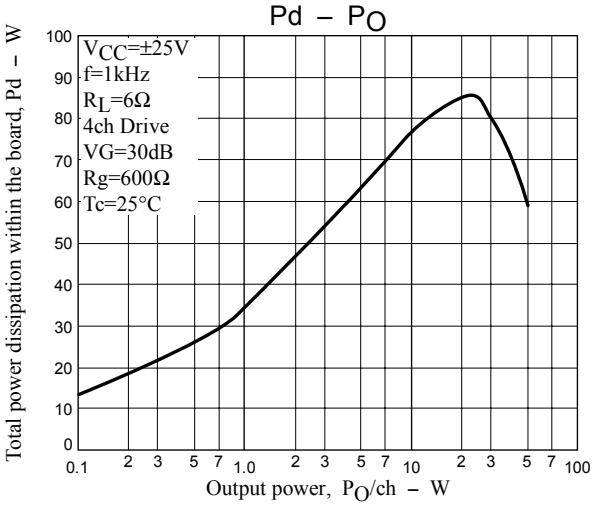
[STK433-000N/-100N/-800Nsr Pin Layout]

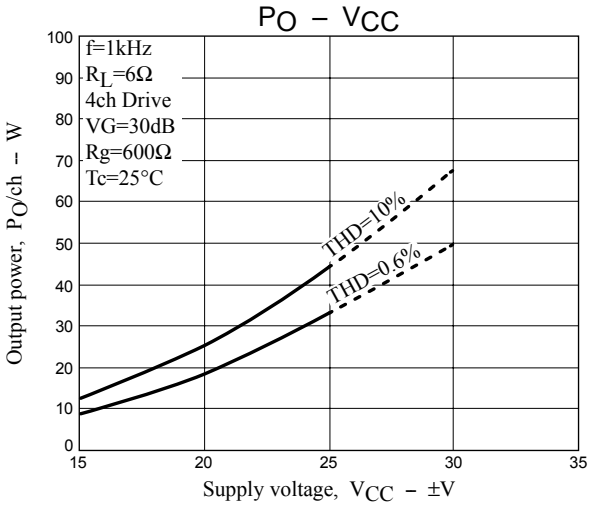
	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15
(Size) 47.0mm×25.6mm×9.0mm	2ch classAB/2.00mm														
STK433-040N 40W/JEITA	-PRE	-VCC	+VCC	OUT/CH1+	OUT/CH1-	OUT/CH2+	OUT/CH2-	+PRE	SUB	GND	IN/CH1	NF/CH1	STAND|BY	NF/CH2	IN/CH2
STK433-060N 50W/JEITA															
(Size) 67.0mm×25.6mm×9.0mm															
STK433-130N 150W/JEITA															

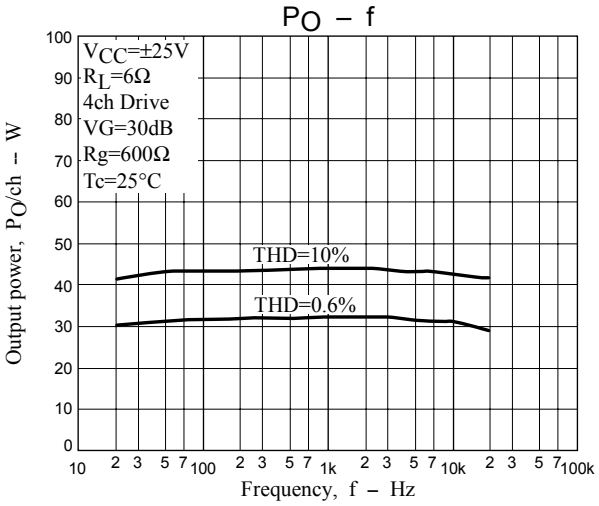
	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	23
(Size) 64.0mm×31.1mm×9.0mm	4ch classAB/2.00mm																						
STK433-840N 40W/JEITA	-PRE	-VCC	+VCC	OUT/CH1+	OUT/CH1-	OUT/CH2+	OUT/CH2-	+PRE	SUB	GND	IN/CH1	NF/CH1	STAND|BY	NF/CH2	IN/CH2	NF/CH3	IN/CH3	IN/CH4	NF/CH4	OUT/CH3-	OUT/CH3+	OUT/CH4-	OUT/CH4+
STK433-870N 60W/JEITA																							
(Size) 78.0mm×44.1mm×9.0mm																							
STK433-890N 80W/JEITA																							

Characteristic of Evaluation Board









A Thermal Design Tip For STK433-840N-E Amplifier

[Thermal Design Conditions]

The thermal resistance (θc-a) of the heat-sink which manages the heat dissipation inside the Hybrid IC will be determined as follow:

(Condition 1) The case temperature (Tc) of the Hybrid IC should not exceed 125°C

$$Pd \times \theta c\text{-}a + Ta < 125°C \cdots\cdots (1)$$

Where Ta : the ambient temperature for the system

(Condition 2) The junction temperature of each power transistor should not exceed 150°C

$$Pd \times \theta c\text{-}a + Pd/N \times \theta j\text{-}c + Ta < 150°C \cdots\cdots (2)$$

Where N : the number of transistors (two for 1 channel , ten for channel)

θj-c : the thermal resistance of each transistor (see specification)

Note that the power consumption of each power transistor is assumed to be equal to the total power dissipation (Pd) divided by the number of transistors (N).

From the formula (1) and (2), we will obtain:

$$\theta c\text{-}a < (125 - Ta)/Pd \cdots\cdots (1)'$$

$$\theta c\text{-}a < (150 - Ta)/Pd - \theta j\text{-}c/N \cdots\cdots (2)'$$

The value which satisfies above formula (1)' and (2)' will be the thermal resistance for a desired heat-sink.

Note that all of the component except power transistors employed in the Hybrid IC comply with above conditions.

[Example of Thermal Design]

Generally, the power consumption of actual music signals are being estimated by the continuous signal of 1/8 P_O max. (Note that the value of 1/8 P_O max may be varied from the country to country.)

(Sample of STK433-840N-E ; 25W×4ch)

If V_{CC} is ±25V, and R_L is 6Ω, then the total power dissipation (Pd) of inside Hybrid IC is as follow;

$Pd = 55.0W$ (at 3.125W output power,1/8 of P_O max)

There are eight (8) transistors in Audio Section of this Hybrid IC, and thermal resistance (θj-c) of each transistor is 4.2°C/W. If the ambient temperature (Ta) is guaranteed for 50°C, then the thermal resistance (θc-a) of a desired heat-sink should be;

From (1)' $\theta c\text{-}a < (125 - 50)/55.0$

< 1.36

From (2)' $\theta c\text{-}a < (150 - 50)/55.0 - 4.2/8$

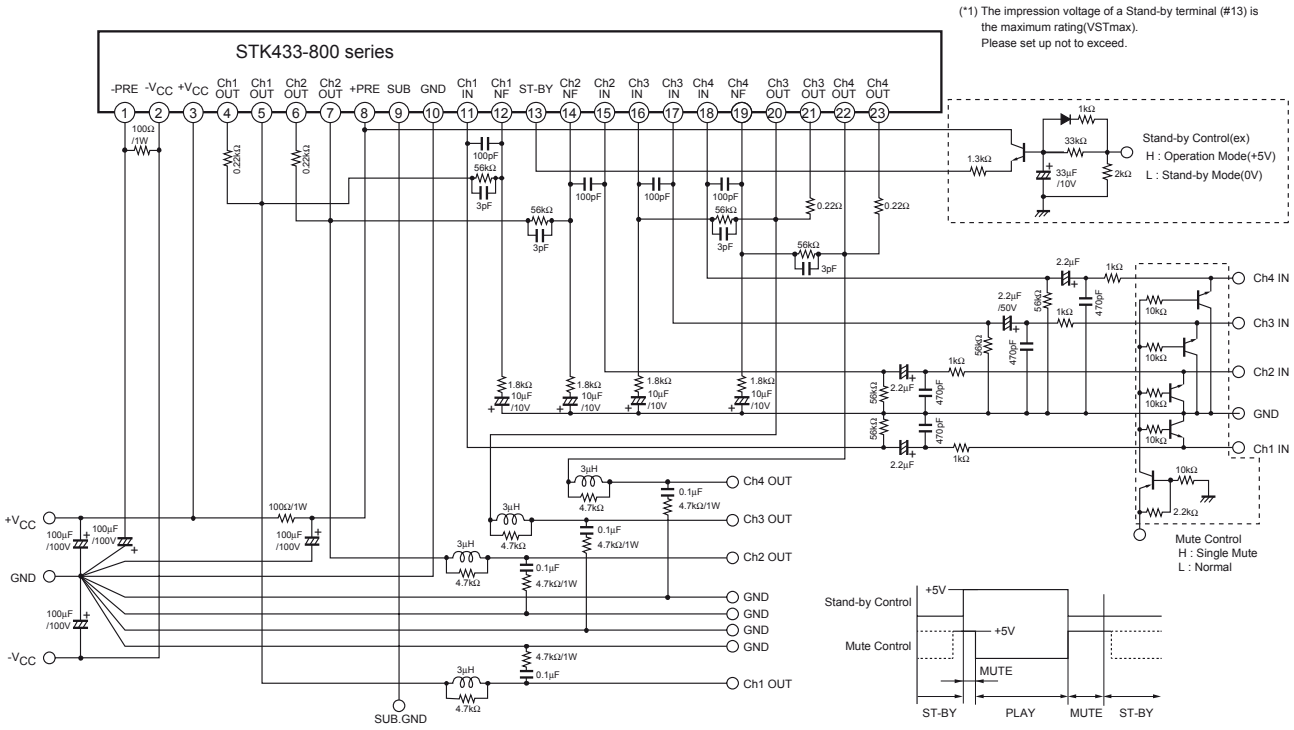
< 1.29

Therefore, in order to satisfy both (1)' and (2)', the thermal resistance of a desired Heat-sink will be 1.29°C/W.

[Note]

Above are reference only. The samples are operated with a constant power supply. Please verify the conditions when your system is actually implemented.

STK433-800 series Stand-by Control & Mute Control & Load-Short Protection Application

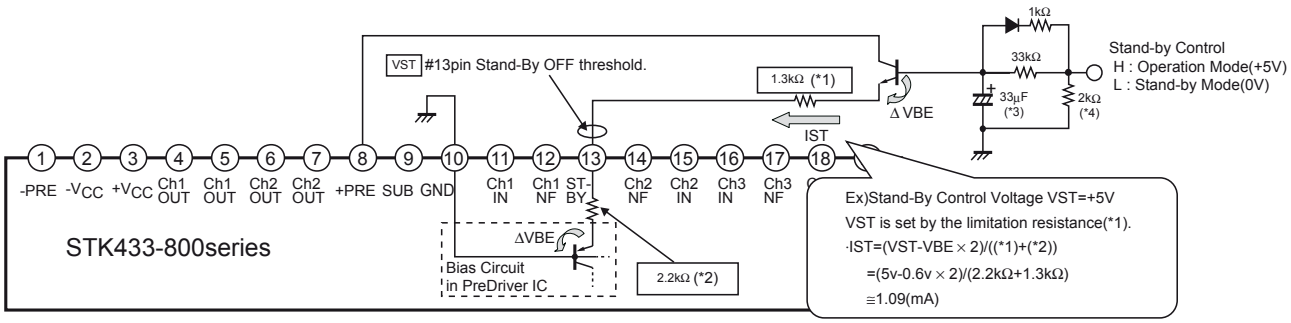


[STK433-800 series Stand-By Control Example]

[Feature]

- The pop noise generated when power supply ON/OFF by using recommendation Stand-By Control Application can be improved.
- Stand-By Control can be done by additionally adjusting the limitation resistance (*1) to the voltage such as Micro computer, the set design is easy.

(Reference circuit) STK433-800 series test circuit To Stand-By Control added +5V.



[Operation explanation]

1) #13pin Stand-By Control Voltage VST

(1) Operation Mode

SW transistor of Stand-By Circuit is turned on when VST ≥ 2.5V or more is impressed, and the power amplifier works.

ex) VST = 3.0V

VST = (*2) × IST+0.6V → 3.0V = 2.2kΩ × IST+0.6V Therefore, IST≅1.09mA

(2) Stand-By Mode

VST ≤ 0.6V or less turns off the SW transistor of Stand-By Circuit by (typ 0V), and the amplifier stops.

ex) VST = 0.6V

VST = (*2) × IST+0.6V → 0.6V = 2.2kΩ × IST+0.6V Therefore, IST≅0mA

(*3) When the power supply is turned on by giving the time constant with the capacitor (*3) when the amplifier works, the pop noise is improved.
(*4) When capacitor (*3) is discharged when the amplifier operation stops, the constant is decided.